



# A Novel High Step-Up Interleaved Converter with Voltage Multiplier Circuit for PV Systems

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**Abstract** – A novel high step-up converter, which is suitable for photovoltaic system is proposed in this project. Through a voltage multiplier module composed of switched capacitors and coupled inductors, the proposed converter obtains high step-up gain. The proposed converter not only reduces the current stress. But also constrains the input current ripple, which decreases the conduction losses and the efficiency gets improved.

**Keywords** - Boost-fly back converter, high step-up, photo-voltaic (PV) system, voltage multiplier module.

## I. INTRODUCTION

Nowadays, renewable energy is increasingly valued and employed worldwide because of energy shortage and environmental contamination [1] – [10]. Renewable energy systems generate low voltage output, and thus, high step-up dc/dc converters have been widely employed in many renewable energy applications such fuel cells, wind power generation, and photovoltaic (PV) systems. Such systems transform energy from renewable sources into electrical energy and convert low voltage into high voltage via a step-up converter, which can convert energy into electricity using a grid-by-grid inverter or dc micro grid. Fig.1 shows a typical renewable energy system that consists of renewable energy sources, a step-up converter, and an inverter for ac application. The high step-up conversion may require two-stage converters with cascade structure for enough step-up gain, which decreases the

efficiency and increases the cost. Thus, a high step-up converter is seen as an important stage in the system because such a system requires a sufficiently high step-up conversion with high efficiency.

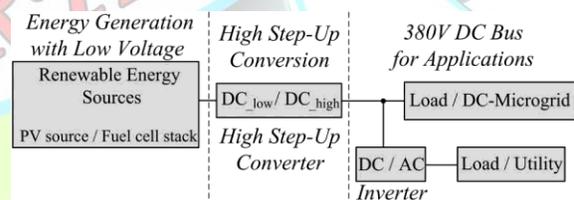


Fig. 1. Typical renewable energy system

Theoretically, conventional step-up converters, such as the boost converter and fly back converter, cannot achieve a high step-up conversion with high efficiency because of the resistances of elements or leakage inductance; also, the voltage stresses are large. Thus, in recent years, many novel high step-up converters have been developed.



Despite these advances, high step-up single-switch converters are unsuitable to operate at heavy load given a large input current ripple, which increases conduction losses.

The conventional interleaved boost converter is an excellent candidate for high-power applications and power factor correction. Unfortunately, the step-up gain is limited, and the voltage stresses on semiconductor components are equal to output voltage. Hence, based on the fore mentioned considerations, modifying a conventional interleaved boost converter for high step-up and high-power application is a suitable approach. To integrate switched capacitors into an interleaved boost converter may make voltage gain reduplicate, but no employment of coupled inductors causes the step-up voltage gain to be limited. Oppositely, to integrate only coupled inductors into an interleaved boost converter may make voltage gain higher and adjustable, but no employment of switched capacitors causes the step-up voltage gain to be ordinary. Thus, the synchronous employment of coupled inductors and switched capacitors is a better concept; moreover, high step-up gain, high efficiency, and low voltage stress are achieved even for high-power applications.

The proposed converter is a conventional interleaved boost converter integrated with a voltage multiplier module, and the voltage multiplier module is composed of switched capacitors and coupled inductors. The coupled inductors can be designed to extend step-up gain, and the switched capacitors offer extra voltage conversion ratio. In addition, when one of the switches

turns off, the energy stored in the magnetizing inductor will transfer via three respective paths; thus, the current distribution not only decreases the conduction losses by lower effective current but also makes currents through some diodes decrease to zero before they turnoff, which all evade diode reverse recovery losses.

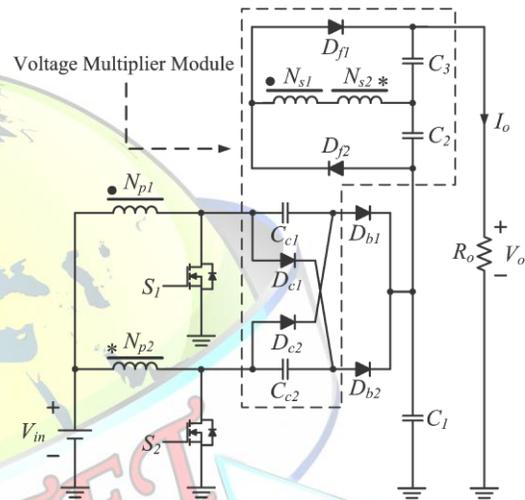


Fig.2. Proposed high step-up converter

## II. ADVANTAGES OF THE PROPOSED CONVERTER

The advantages of the proposed converter are as follows.

- 1) The proposed converter increases the life time of renewable energy sources and makes it suitable for high-power applications.
- 2) Due to the lossless passive clamp performance, leakage energy is recycled to the output terminal. Hence, large voltage spikes across the main switches are reduced and the efficiency is improved.



3) The converter achieves the high step-up gain that renewable energy systems require.

4) Low cost and high efficiency are achieved by employment of the low-voltage-rated power switch with low  $R_{DS}$  (ON); also, the voltage stresses on main switches and diodes are substantially lower than output voltage.

5) The inherent configuration of the proposed converter makes some diodes decrease conduction losses and all eviate diode reverse recovery losses.

### III. OPERATING PRINCIPLES

The proposed high step-up inter leaved converter with a voltage multiplier module is shown in Fig.2. The voltage multiplier module is composed of two coupled inductors and two switched capacitors and is inserted between a conventional interleaved boost converter to form a modified boost–flyback–forward inter leaved structure. When the switches turn off by turn, the phase whose switch is in OFF state performs as a flyback converter, and the other phase whose switch is in ON state performs as a forward converter.

Primary windings of the coupled inductors with  $N_p$  turns are employed to decrease input current ripple, and secondary windings of the coupled inductors with  $N_s$  turns are connected in series to extend voltage gain. The turn ratios of the coupled inductors are the same. The coupling references of the inductors are denoted by “.” and “\*”.

The equivalent circuit of the proposed converter is shown in Fig.3, where  $L_{m1}$  and  $L_{m2}$  are the magnetizing inductors;  $L_{k1}$  and  $L_{k2}$  represent the leakage inductors;  $L_s$  represents the series leakage inductors in the secondary side;  $S_1$  and  $S_2$  denote the power switches;  $C_{c1}$  and  $C_{c2}$  are the switched capacitors; and  $C_1$ ,  $C_2$ , and  $C_3$  are the output capacitors.  $D_{c1}$  and  $D_{c2}$  are the clamp diodes,  $D_{b1}$  and  $D_{b2}$  represent the output diodes for boost operation with switched capacitors,  $D_{f1}$  and  $D_{f2}$  represent the output diodes for flyback–forward operation and  $n$  is defined as turn ratio  $N_s/N_p$ .

In the circuit analysis, the proposed converter operates in continuous conduction mode (CCM), and the duty cycles of the power switches during steady operation are greater than 0.5 and are interleaved with a  $180^\circ$  phase shift. The key steady waveform in one switching period of the proposed converter contains six modes, which are depicted in Fig.4, and Fig.5 shows the topological stages of the circuit.

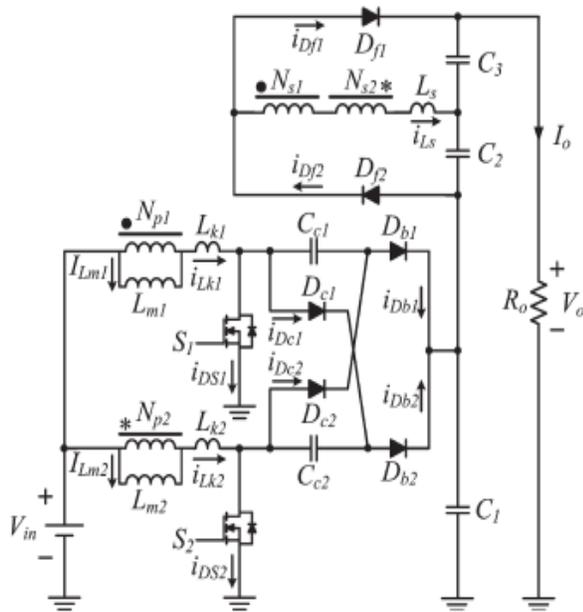


Fig.3. Equivalent circuit of the proposed converter

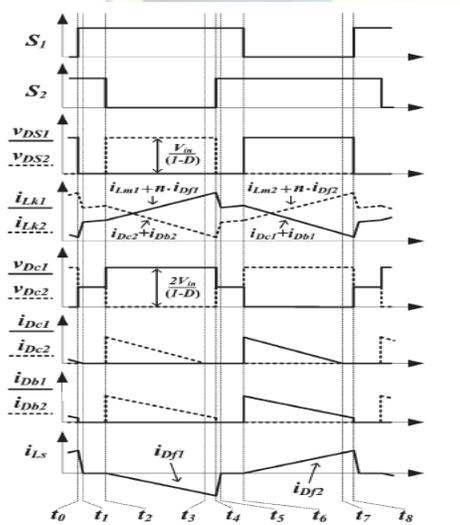


Fig.4. Steady waveform of the proposed converter in CCM

#### IV. STEADY-STATE ANALYSIS

The transient characteristics of circuitry are disregarded to simplify the circuit performance analysis of the proposed converter in CCM, and some formulated assumptions are as follows.

- 1) All of the components in the proposed converter are ideal.
- 2) Leakage inductors  $L_{k1}$ ,  $L_{k2}$ , and  $L_s$  are neglected.
- 3) Voltages on all capacitors are considered to be constant because of infinitely large capacitance.
- 4) Due to the completely symmetrical interleaved structure, the related components are defined as the corresponding symbols such as  $D_{c1}$  and  $D_{c2}$  defined as  $D_c$ .

#### A. Step-Up Gain

The voltage on clamp capacitor  $C_c$  can be regarded as an output voltage of the boost converter; thus, voltage  $V_{C_c}$  can be derived from the equation

$$V_{C_c} = \frac{1}{1-D} V_{in} \quad (1)$$

When one of the switches turns off, voltage  $V_{C_1}$  can obtain a double output voltage of the boost converter derived from

$$V_{C_1} = \frac{1}{1-D} V_{in} + V_{C_c} = \frac{2}{1-D} V_{in} \quad (2)$$

The outputs filter capacitors  $C_2$  and  $C_3$  are charged by energy transformation from the primary side. When  $S_2$  is in ON state and  $S_1$  is in OFF state,  $V_{C_2}$  is equal to the sum of the



induced voltage of  $N_{S1}$  and the induced voltage of  $N_{S2}$ , and when  $S_{1}$  is in ON state and  $S_{2}$  is in OFF state,  $V_{C3}$  is also equal to the sum of the induced voltage of  $N_{S1}$  and the induced voltage of  $N_{S2}$ . Thus, voltages  $V_{C2}$  and  $V_{C3}$  can be derived from

$$V_{C2} = V_{C3} = n V_{in} \left( 1 + \frac{D}{1-D} \right) = \frac{n}{1-D} V_{in} \quad (3)$$

This equation represents low-voltage-rated MOSFET with low  $R_{DS(ON)}$  can be adopted for the proposed converter to reduce conduction losses and costs. These voltage stresses can be derived from

$$V_{Dc1} = V_{Dc2} = \frac{2}{1-D} V_{in} = \frac{1}{n+1} V_0 \quad (7)$$

$$V_{Db1} = V_{Db2} = V_{C1} - V_{C2} = \frac{2}{1-D} V_{in} = \frac{1}{2n+2} V_0 \quad (8)$$

The output voltage can be derived from

$$V_0 = V_{C1} + V_{C2} + V_{C3} = \frac{2n+2}{1-D} V_{in} \quad (4)$$

Although the voltage stress on the diode  $D_f$  increases as the turn ratio  $n$  increases, the voltage stress on the diodes  $D_f$  is always lower than the output voltage.

In addition, the voltage gain of the proposed converter is

$$\frac{V_0}{V_{in}} = \frac{2n+2}{1-D} \quad (5)$$

## V. DESIGN AND EXPERIMENT OF PROPOSED CONVERTER

### B. VOLTAGE STRESS ON SEMI CONDUCTOR COMPONENT

Voltage ripple on capacitor was ignored to simplify the voltage stress analysis of the components of the proposed converter. The voltage stress on power switch  $S$  was clamped and derived from

$$V_{S1} = V_{S2} = \frac{2}{1-D} V_{in} = \frac{1}{2n+2} V_0 \quad (6)$$

A 1-kW prototype of the proposed high step-up converter is tested. The electrical specifications are  $V_{in} = 40$  V,  $V_0 = 380$  V, and  $f_s = 40$  kHz. The major components have been chosen as follows: Magnetizing inductors  $L_{m1}$  and  $L_{m2} = 133$   $\mu$ H; turn ratio  $n = 1$ ; power switches  $S1$  and  $S2$  are IRFP4227; diodes  $D_{c1}$  and  $D_{c2}$  are BYQ28E-200; diodes  $D_{b1}$ ,  $D_{b2}$ ,  $D_{f1}$ , and  $D_{f2}$  are FCF06A-40; capacitors  $C_{c1}$ ,  $C_{c2}$ ,  $C_2$ , and  $C_3 = 220$   $\mu$ F; and  $C_1 = 470$   $\mu$ F.

The design consideration of the proposed converter includes component selection and coupled inductor design, which are based on the analysis presented in the previous section. In the proposed converter, the values of the primary leakage inductors of the coupled inductors are set as close as possible for current sharing performance, and the leakage

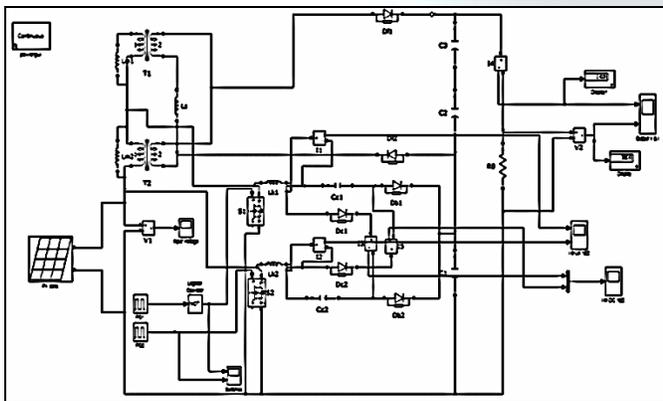


inductors  $Lk1$  and  $Lk2$  are  $1.6 \mu\text{H}$ . Due to the performances of high step-up gain, the turn ratio  $n$  can be set as one for the prototype circuit with 40-V input voltage and 380-V output to reduce cost, volume, and conduction loss of the winding. Thus, the copper resistances which affect efficiency much can be decreased.

#### V. DESIGN PARAMETERS

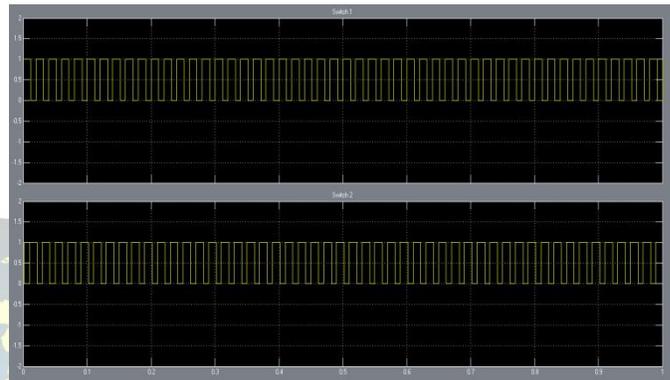
PARAMETER	VALUE
Input Voltage	40 V DC
Output Voltage	380 V DC
Switching Frequency	40 KHz
Magnetizing Inductors ( $Lm1$ & $Lm2$ )	$133 \mu\text{H}$
Capacitors ( $Cc1$ , $Cc2$ , $C2$ & $C3$ )	$220 \mu\text{F}$
Capacitor ( $C1$ )	$470 \mu\text{F}$
Output Power	1000 W

#### VI. MATLAB SIMULATION DIAGRAM



#### VII. WAVEFORMS

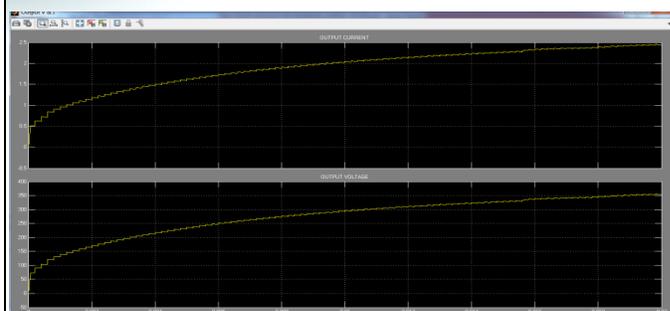
a) Switching Pulses:



b) Current Through Leakage Inductances:



c) Output Current and Output Voltage:





## VIII CONCLUSION

This paper has presented the theoretical analysis of steady state, related consideration, simulation results, and experimental results for the proposed converter. The proposed converter has successfully implemented an efficient high step-up conversion through the voltage multiplier module. The interleaved structure reduces the input current ripple and distributes the current through each component. In addition, the lossless passive clamp function recycles the leakage energy and constrains a large voltage spike across the power switch. Meanwhile, the voltage stress on the power switch is restricted and much lower than the output voltage (380V). Furthermore, the full-load efficiency is 96.4% at  $P_o=1000W$ , and the highest efficiency is 97.1% at  $P_o=400W$ . Thus, the proposed converter is suitable for high-power or renewable energy applications that need high step-up conversion.

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